

Necessary and Sufficient Conditions for Input-Output Finite-Time Stability of Impulsive Dynamical Systems

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Outline

1 Motivations

2 Preliminaries

- Notation

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Input-output finite-time stability vs classic IO stability

IO stability

A dynamical system is said to be IO \mathcal{L}_p -stable if for any input of class \mathcal{L}_p , the system exhibits a corresponding output which belongs to the same class

IO-FTS

A dynamical system is defined to be IO-FTS if, given a class of norm bounded input signals over a specified time interval T , the outputs of the system do not exceed an assigned threshold during T

Main features of IO-FTS

IO-FTS:

- involves signals defined over a finite time interval
- does not necessarily require the inputs and outputs to belong to the same class
- specifies a *quantitative* bounds on both inputs and outputs

IO stability and IO-FTS are independent concepts

Examples of application



G. Ambrosino, M. Ariola, G. De Tommasi, A. Pironti
Plasma Vertical Stabilization in the ITER Tokamak via
Constrained Static Output Feedback
IEEE Trans. Contr. Tech., 2011



F. Amato, G. Carannante, G. De Tommasi, A. Pironti
Input-Output Finite-Time Stabilization of Linear Systems with
Input Constraints
IET Contr. Theory Appl., 2014

Contribution of the paper

- In this paper we show that the sufficient condition to check IO-FTS of time-dependent Impulsive Dynamical Linear Systems (IDLS), which is expressed in terms of a coupled difference/differential LMI (D/DLMI) feasibility problem, and which was originally given in



F. Amato, G. Carannante, G. De Tommasi

Input-output Finite-Time Stabilisation of a class of Hybrid Systems via Static Output Feedback

Int. J. Contr., 2011

is also **necessary**.

- An alternative, **and numerically more efficient**, necessary and sufficient condition for IO-FTS is proved, which requires the solution of a coupled difference/differential Lyapunov equation (D/DLE).

Notation

- \mathcal{L}_p denotes the space of vector-valued signals whose p -th power is absolutely integrable over $[0, +\infty)$.
- The restriction of \mathcal{L}_p to the time interval $\Omega := [t_0, t_0 + T]$ is denoted by $\mathcal{L}_p(\Omega)$.
- Given the time interval Ω , a symmetric positive definite matrix-valued function $R(\cdot)$, bounded on Ω , and a vector-valued signal $s(\cdot) \in \mathcal{L}_p(\Omega)$, the weighted signal norm

$$\left(\int_{\Omega} [s^T(\tau)R(\tau)s(\tau)]^{\frac{p}{2}} d\tau \right)^{\frac{1}{p}},$$

will be denoted by $\|s(\cdot)\|_{p,R}$. If $p = \infty$

$$\|s(\cdot)\|_{\infty,R} = \operatorname{ess\,sup}_{t \in \Omega} [s^T(t)R(t)s(t)]^{\frac{1}{2}}.$$

Impulsive Dynamical Linear Systems

The class of time dependent Impulsive Dynamical Linear Systems is described by

$$\Gamma : \begin{cases} \dot{x}(t) = A(t)x(t) + G(t)w(t), & x(t_0) = 0, & t \notin \mathcal{T} & (1a) \\ x^+(t_i) = J(t_i)x(t_i), & t_i \in \mathcal{T} & & (1b) \\ y(t) = C(t)x(t), & \forall t, & & (1c) \end{cases}$$

- $J(\cdot)$ is the matrix-valued function that describes the *resetting law* of the system.
- The elements of the set $\mathcal{T} = \{t_1, t_2, \dots\}$ are called *resetting times*.
- According to the continuous-time dynamics (1a) and the resetting law (1b), an IDLS presents a left-continuous trajectory with a finite jump from $x(t_i)$ to $x^+(t_i)$ at each resetting time $t_i \in \mathcal{T}$.
- Being interested in the dynamic behaviour of the IDLS in the time interval Ω , the number of resetting times in Ω is assumed equal to N .
- It is also assumed that the first resetting time $t_1 \in \mathcal{T}$ is such that $t_1 > t_0$.

The state transition matrix $\Phi(t, \tau)$ of an IDLS

The following properties for the transition matrix $\Phi(t, \tau)$ of (1) hold

$$\Phi(t_0, t_0) = I, \quad (2a)$$

$$\frac{\partial}{\partial t} \Phi(t, t_0) = A(t)\Phi(t, t_0), \quad t \notin \mathcal{T} \quad (2b)$$

$$\Phi^+(t_i, t_0) = J(t_i)\Phi(t_i, t_0), \quad t_i \in \mathcal{T}. \quad (2c)$$

Given $t_k < t < t_{k+1}$, with $t_k, t_{k+1} \in \mathcal{T}$, it is (see also [Medina 2007](#))

$$\Phi(t, t_0) = \Phi_{k+1}(t, t_k)J(t_k)\Phi_k(t_k, t_{k-1})J(t_{k-1}) \cdots J(t_2)\Phi_2(t_2, t_1)J(t_1)\Phi_1(t_1, t_0), \quad (3)$$

where for $j = 1, \dots, N$, $\Phi_j(\cdot, \cdot)$ satisfies

$$\frac{\partial}{\partial t} \Phi_j(t, t_{j-1}) = A(t)\Phi_j(t, t_{j-1}), \quad t \in [t_{j-1}, t_j[, \quad \Phi_j(t_{j-1}, t_{j-1}) = I,$$

and $\Phi_{N+1}(\cdot, \cdot)$,

$$\frac{\partial}{\partial t} \Phi_{N+1}(t, t_N) = A(t)\Phi_{N+1}(t, t_N), \quad t \in [t_N, t_0 + T], \quad \Phi_{N+1}(t_N, t_N) = I.$$

Given (3), it is straightforward to verify that the **impulsive response** of (1), is given by

$$H(t, \tau) = C(t)\Phi(t, \tau)G(\tau)\delta_{-1}(t - \tau).$$

Reachability Gramian of IDLSs

- Also the **reachability Gramian** $W_r(\cdot, \cdot)$ of an IDLS can be recursively defined
- In Medina and Lawrence 2009 it has been shown that $W_r(\cdot, \cdot)$ is the unique symmetric and positive semidefinite solution of the following D/DLE

$$\dot{W}_r(t, t_0) = A(t)W_r(t, t_0) + W_r(t, t_0)A^T(t) + G(t)G^T(t), \quad t \notin \mathcal{T} \quad (4a)$$

$$W_r^+(t_i, t_0) = J(t_i)W_r(t_i, t_0)J^T(t_i), \quad t_i \in \mathcal{T} \quad (4b)$$

$$W_r(t_0, t_0) = 0 \quad (4c)$$

Formal definition of IO-FTS for IDLSs

IO-FTS of IDLSs

Given a positive scalar T , a class of input signals \mathcal{W} defined over $\Omega = [t_0, t_0 + T]$, a continuous, positive definite matrix-valued function $Q(\cdot)$ defined in Ω , system (1) is said to be IO-FTS with respect to $(\mathcal{W}, Q(\cdot), \Omega)$ if

$$w(\cdot) \in \mathcal{W} \Rightarrow y^T(t)Q(t)y(t) < 1, \quad \forall t \in \Omega.$$

Class square integrable disturbances - \mathcal{W}_2

$$\mathcal{W}_2(\Omega, R(\cdot)) := \{w(\cdot) \in \mathcal{L}_2(\Omega) : \|w\|_{2,R} \leq 1\}.$$

IDLSs as linear operators

- The IDLS (1) can be regarded as a **linear operator**

$$\Gamma : w(\cdot) \in \mathcal{L}_2(\Omega) \mapsto y(\cdot) \in \mathcal{L}_\infty(\Omega), \quad (5)$$

- Equipping the $\mathcal{L}_2(\Omega)$ and $\mathcal{L}_\infty(\Omega)$ spaces with the weighted norms $\|\cdot\|_{2,R}$ and $\|\cdot\|_{\infty,Q}$, respectively, the induced norm of (5) is equal to

$$\|\Gamma\| = \sup_{\|w(\cdot)\|_{2,R}=1} \left[\|y(\cdot)\|_{\infty,Q} \right].$$

Theorem 1

Given a time interval Ω , the class of input signals \mathcal{W}_2 , and a continuous positive definite matrix-valued function $Q(\cdot)$, system (1) is IO-FTS with respect to $(\mathcal{W}_2, Q(\cdot), \Omega)$ *if and only if* $\|\Gamma\| < 1$.

Norm of the linear operator Γ

Theorem 2

Given the IDLS (1), the norm of the corresponding linear operator (5) is given by

$$\|\Gamma\| = \operatorname{ess\,sup}_{t \in \Omega} \lambda_{\max}^{\frac{1}{2}} \left(Q^{\frac{1}{2}}(t) C(t) W(t) C^T(t) Q^{\frac{1}{2}}(t) \right), \quad (6)$$

for all $t \in \Omega$; $W(\cdot)$ is the piecewise continuously differentiable positive semidefinite matrix-valued solution of

$$\dot{W}(t) = A(t)W(t) + W(t)A^T(t) + G(t)R(t)^{-1}G^T(t), \quad t \notin \mathcal{T} \quad (7a)$$

$$W^+(t_i) = J(t_i)W(t_i)J^T(t_i), \quad t_i \in \mathcal{T} \quad (7b)$$

$$W(t_0) = 0. \quad (7c)$$

Main result

Theorem 3

The following statements are equivalent:

- i) The IDLS (1) is IO finite-time stable with respect to $(\mathcal{W}_2, Q(\cdot), \Omega)$.
- ii) The inequality $\text{ess sup}_{t \in \Omega} \lambda_{\max}(Q^{\frac{1}{2}}(t)C(t)W(t)C^T(t)Q^{\frac{1}{2}}(t)) < 1$ holds, where $W(\cdot)$ is the solution of (7).
- iii) The coupled D/DLMI

$$\begin{pmatrix} \dot{P}(t) + A^T(t)P(t) + P(t)A(t) & P(t)G(t) \\ G^T(t)P(t) & -R(t) \end{pmatrix} < 0$$

$$t \notin \mathcal{T}, \quad (8a)$$

$$J^T(t_i)P^+(t_i)J(t_i) - P(t_i) < 0, \quad t \in \mathcal{T}, \quad (8b)$$

$$P(t) > C^T(t)Q(t)C(t), \quad t \in \Omega \quad (8c)$$

admits a piecewise continuously differentiable positive definite solution $P(\cdot)$ over Ω .

Comparison of the computational burden 1/3

Let us consider the time-varying IDLS

$$\begin{aligned} A &= \begin{pmatrix} -2.5 + 0.2 \cdot t & -6.3 \\ 4 & 0.2 \cdot t \end{pmatrix}, & G &= \begin{pmatrix} 2 \\ 0 \end{pmatrix}, \\ C &= (1.2 \quad 3.2), & J &= \begin{pmatrix} 1.1 & 0 \\ 0 & -0.8 \end{pmatrix}. \end{aligned} \quad (9)$$

The time interval we consider in this example is $\Omega = [0, 2]$, while the resetting times are

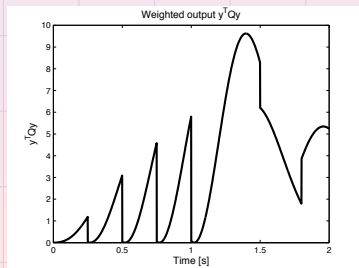
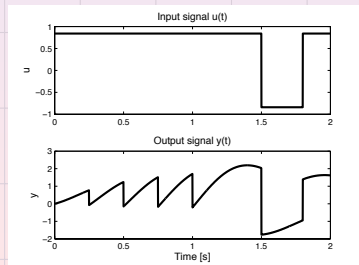
$$\mathcal{T} = \{0.25, 0.5, 0.75, 1, 1.5, 1.8\}.$$

The input weighting matrix is taken constant and equal to

$$R = 0.7.$$

Comparison of the computational burden 2/3

Time response of the IDLS (9) in the interval Ω when an input in $\mathcal{W}_2([0, 2], 0.7)$ is considered, and when **Q is taken equal to 2**.



Comparison of the computational burden 3/3

Table: Values of Q_{\max} obtained exploiting condition **ii)** in Theorem 3 for the IDLS system (9)

Sample Time (T_s) [ms]	Q_{\max}	Computation time for the solution of the D/DLE (7) [s]
10	0.0900	0.19
1	0.0910	0.22
0.1	0.0918	0.7

Table: Values of Q_{\max} obtained exploiting condition **iii)** in Theorem 3 for the IDLS system (9)

Sample Time (T_s) [ms]	Q_{\max}	Average computation time for a single iteration [s]
50	0.0740	2.6
25	0.0796	14.2
10	0.0837	298.4

Conclusions

- Necessary and sufficient conditions for IO-FTS of IDLSs have been presented for the class of \mathcal{W}_2 disturbances
- The D/DLMI formulation can be extended to solve the IO finite-time stabilization problem either via state-feedback, or via output-feedback

Thank you!