

Necessary and Sufficient Conditions for Input-Output Finite-Time Stability of Linear Time-Varying Systems

Francesco Amato¹ Giuseppe Carannante²
Gianmaria De Tommasi² Alfredo Pironti²

¹Università degli Studi Magna Græcia di Catanzaro, Catanzaro, Italy,

²Università degli Studi di Napoli Federico II, Napoli, Italy

Joint 50th IEEE Conference on Decision and Control &
European Control Conference
December 12–15, 2011, Orlando, Florida

Outline

- 1 Motivations**
- 2 Preliminaries**
 - Notation
 - Problem Statement
 - Preliminary result
- 3 Main Theorem**
- 4 Numerical example**

Input-output finite-time stability vs classic IO stability

IO stability

A system is said to be IO \mathcal{L}_p -stable if for any input of class \mathcal{L}_p , the system exhibits a corresponding output which belongs to the same class

IO-FTS

A system is defined to be IO-FTS if, given a class of norm bounded input signals over a specified time interval T , the outputs of the system do not exceed an assigned threshold during T

Main features of IO-FTS

IO-FTS:

- involves signals defined over a finite time interval
- does not necessarily require the inputs and outputs to belong to the same class
- specifies a *quantitative* bounds on both inputs and outputs

IO stability and IO-FTS are independent concepts

Contribution of the paper

- In this paper we show that, in the case of \mathcal{L}_2 inputs, the sufficient condition given in



F. Amato, R. Ambrosino, G. De Tommasi, C. Cosentino
Input-output finite-time stabilization of linear systems
Automatica, 2010

is also **necessary**.

- To prove this result, a machinery involving the *teachability gramian* is used.

Notation

- \mathcal{L}_p denotes the space of vector-valued signals whose p -th power is absolutely integrable over $[0, +\infty)$.
- The restriction of \mathcal{L}_p to $\Omega := [t_0, t_0 + T]$ is denoted by $\mathcal{L}_p(\Omega)$.
- Given the time interval Ω , a symmetric positive definite matrix-valued function $R(\cdot)$, bounded on Ω , and a vector-valued signal $s(\cdot) \in \mathcal{L}_p(\Omega)$, the weighted signal norm

$$\left(\int_{\Omega} [s^T(\tau)R(\tau)s(\tau)]^{\frac{p}{2}} d\tau \right)^{\frac{1}{p}},$$

will be denoted by $\|s(\cdot)\|_{p,R}$. If $p = \infty$

$$\|s(\cdot)\|_{\infty,R} = \operatorname{ess\,sup}_{t \in \Omega} [s^T(t)R(t)s(t)]^{\frac{1}{2}}.$$

LTV systems as Linear Operator

Let us consider a LTV system in the form

$$\Gamma : \begin{cases} \dot{x}(t) = A(t)x(t) + G(t)w(t), & x(t_0) = 0 \\ y(t) = C(t)x(t) \end{cases} \quad (1)$$

Γ can be viewed as a linear operator mapping input signals ($w(\cdot)$'s) into output signals ($y(\cdot)$'s).

$\Phi(t, \tau)$ denotes the state transition matrix of system (1).

Reachability Gramian

The *reachability Gramian* of system (1) is defined as

$$W_r(t, t_0) \triangleq \int_{t_0}^t \Phi(t, \tau) G(\tau) G^T(\tau) \Phi^T(t, \tau) d\tau.$$

$W_r(t, t_0)$ is symmetric and positive semidefinite for all $t \geq t_0$.

Given system (1), $W_r(t, t_0)$ is the unique solution of the matrix differential equation

$$\dot{W}_r(t, t_0) = A(t)W_r(t, t_0) + W_r(t, t_0)A^T(t) + G(t)G^T(t), \quad (2a)$$

$$W_r(t_0, t_0) = 0 \quad (2b)$$

IO-FTS of LTV systems

Given a positive scalar T , a class of input signals \mathcal{W} defined over $\Omega = [t_0, t_0 + T]$, a positive definite matrix-valued function $Q(\cdot)$ defined in Ω , system (1) is said to be IO-FTS with respect to $(\mathcal{W}, Q(\cdot), \Omega)$ if

$$w(\cdot) \in \mathcal{W} \Rightarrow y^T(t)Q(t)y(t) < 1, \quad t \in \Omega.$$

In this work we consider the class of norm bounded square integrable signals over Ω

$$\mathcal{W}_2(\Omega, R(\cdot)) := \{w(\cdot) \in \mathcal{L}_2(\Omega) : \|w\|_{2,R} \leq 1\},$$

where $R(\cdot)$ denotes a continuous positive definite matrix-valued function.

Linear operator

The LTV system (1) is regarded as a linear operator that maps signals from the space $\mathcal{L}_2(\Omega)$ to the space $\mathcal{L}_\infty(\Omega)$

$$\Gamma : w(\cdot) \in \mathcal{L}_2(\Omega) \mapsto y(\cdot) \in \mathcal{L}_\infty(\Omega). \quad (3)$$

If we equip the $\mathcal{L}_2(\Omega)$ and $\mathcal{L}_\infty(\Omega)$ spaces with the weighted norms $\|\cdot\|_{2,R}$ and $\|\cdot\|_{\infty,Q}$, respectively, the induced norm of the linear operator (3) is given by

$$\|\Gamma\| = \sup_{\|w(\cdot)\|_{2,R}=1} \left[\|y(\cdot)\|_{\infty,Q} \right],$$

Theorem 1

Given a time interval Ω , the class of input signals \mathcal{W}_2 , and a continuous positive definite matrix-valued function $Q(\cdot)$, system (1) is IO-FTS with respect to $(\mathcal{W}_2, Q(\cdot), \Omega)$ if and only if $\|\Gamma\| < 1$.

Dual operator

Given the linear operator (3), its **dual operator** is

$$\bar{\Gamma} : z(\cdot) \in \mathcal{L}_1(\Omega) \mapsto v(\cdot) \in \mathcal{L}_2(\Omega),$$

with

$$\|\bar{\Gamma}\| = \sup_{\|z(\cdot)\|_{1,Q}=1} \left[\|v(\cdot)\|_{2,R} \right].$$

By definition it holds

$$\|\Gamma\| = \|\bar{\Gamma}\|, \quad (4)$$

and

$$\langle z, \Gamma w \rangle = \langle \bar{\Gamma} z, w \rangle, \quad (5)$$

where $z(\cdot) \in \mathcal{L}_1(\Omega)$ and $w(\cdot) \in \mathcal{L}_2(\Omega)$.

Theorem 2

Given the LTV system (1), the norm of the corresponding linear operator (3) is given by

$$\|\Gamma\| = \operatorname{ess\,sup}_{t \in \Omega} \lambda_{\max}^{\frac{1}{2}} \left(Q^{\frac{1}{2}}(t) C(t) W(t, t_0) C^T(t) Q^{\frac{1}{2}}(t) \right), \quad (6)$$

for all $t \in \Omega$, where $\lambda_{\max}(\cdot)$ denotes the maximum eigenvalue, and $W(t, t_0)$ is the positive semidefinite matrix-valued solution of

$$\begin{aligned} \dot{W}(t, t_0) &= A(t)W(t, t_0) + W(t, t_0)A^T(t) \\ &\quad + G(t)R(t)^{-1}G^T(t) \end{aligned} \quad (7a)$$

$$W(t_0, t_0) = 0 \quad (7b)$$

Sketch of proof - 1

- For the sake of simplicity, the weighting matrices $R(t)$ and $Q(t)$ are set equal to the identity; it follows that the solution of (7) is given by the reachability gramian $W_r(t, t_0)$;
- Considering the dual operator $\bar{\Gamma}$, proving (6) is equivalent to show

$$\|\bar{\Gamma}\| = \operatorname{ess\,sup}_{t \in \Omega} \lambda_{\max}^{\frac{1}{2}} \left(C(t) W_r(t, t_0) C^T(t) \right).$$

- We denote with

$$\bar{H}(t, \tau) = G^T(t) \Phi^T(\tau, t) C^T(\tau) \delta_{-1}(\tau - t)$$

the impulsive response of the **dual system**

$$\bar{\Gamma} : \begin{cases} \dot{\tilde{x}}(t) = -A^T(t) \tilde{x}(t) - C^T(t) z(t) \\ v(t) = G^T(t) \tilde{x}(t) \end{cases} .$$

Sketch of proof - 2

- Using $\bar{H}(t, \tau)$ it is possible to show that

$$\|v(\cdot)\|_2 = \left\| \int_{\Omega} \bar{H}(\cdot, \tau) z(\tau) d\tau \right\|_2 \leq \operatorname{ess\,sup}_{t \in \Omega} \lambda_{\max}^{\frac{1}{2}} \left(C(t) W_r(t, t_0) C^T(t) \right) \cdot \|z(\cdot)\|_1$$

- Hence

$$\|\bar{\Gamma}\| \leq \operatorname{ess\,sup}_{t \in \Omega} \lambda_{\max}^{\frac{1}{2}} \left(C(t) W_r(t, t_0) C^T(t) \right)$$

- Exploiting similar arguments as in



D. A. Wilson

Convolution and hankel operator norms for linear

IEEE Trans. on Auto. Contr., 1989

it is possible to show that

$$\|\bar{\Gamma}\| = \operatorname{ess\,sup}_{t \in \Omega} \lambda_{\max}^{\frac{1}{2}} \left(C(t) W_r(t, t_0) C^T(t) \right),$$

which proves the theorem.

Remark

If the system matrices in (1) and the weighting matrices $R(\cdot)$ and $Q(\cdot)$ are assumed to be continuous, in the closed time interval Ω the condition (6) is equivalent to

$$\|\Gamma\| = \max_{t \in \Omega} \lambda_{\max}^{\frac{1}{2}} \left(Q^{\frac{1}{2}}(t) C(t) W(t, t_0) C^T(t) Q^{\frac{1}{2}}(t) \right).$$

Theorem 3

The following statements are equivalent:

- i) System (1) is IO-FTS with respect to $(\mathcal{W}_2, Q(\cdot), \Omega)$.
- ii) The inequality

$$\lambda_{\max}(Q^{\frac{1}{2}}(t)C(t)W(t, t_0)C^T(t)Q^{\frac{1}{2}}(t)) < 1 \quad (8)$$

holds for all $t \in \Omega$, where $W(\cdot, \cdot)$ is the positive semidefinite solution of the Differential Lyapunov Equality (DLE) (7).

- iii) The coupled DLMI/LMI

$$\begin{pmatrix} \dot{P}(t) + A^T(t)P(t) + P(t)A(t) & P(t)G(t) \\ G^T(t)P(t) & -R(t) \end{pmatrix} < 0 \quad (9a)$$

$$P(t) > C^T(t)Q(t)C(t), \quad (9b)$$

admits a positive definite solution $P(\cdot)$ over Ω .

Sketch of proof

- The equivalence of the three statements is proved by showing that **i) \Rightarrow ii)**, **ii) \Rightarrow iii)**, and **iii) \Rightarrow i)**.
- A technical lemma is exploited to show that solving the DLE is equivalent to solve a matrix inequality.

Comparison

- The conditions stated in Theorem 3 are all necessary and sufficient.
- The numerical implementation of such conditions introduces some conservativeness.
- In order to compare each other, from the computational point of view the output weighting matrix is left as a free parameter.
- We define Q_{max} as the maximum value of the matrix Q such that a system is IO-FTS.
- To recast the DLMI condition (9) in terms of LMIs, the matrix-valued functions $P(\cdot)$ has been assumed piecewise linear. In particular, the time interval Ω has been divided in $n = T/T_s$ subintervals.

Results

In the paper we have considered the system

$$A(t) = \begin{pmatrix} 0.5 + t & 0.1 \\ 0.4 & -0.3 + t \end{pmatrix}, G = \begin{pmatrix} 1 \\ 1 \end{pmatrix}, C = (1 \ 1),$$

together with the following IO-FTS parameters:

$$R = 1, \Omega = [0, 0.5].$$

Maximum values of Q satisfying Theorem 3. The results have been obtained by using a PC equipped with an Intel i7-720QM processor and 4 GB of RAM.

IO-FTS condition	Sample Time (T_s)	Estimate of Q_{max}	Computation time [s]
DLMI (9)	0.05	0.2	2.5
	0.025	0.25	12.7
	0.0125	0.29	257
	0.00833	0.3	1259
Solution of (7) and inequality (8)	0.003	0.345	6

Conclusions

- Necessary and sufficient conditions for IO-FTS have been presented in this paper for the class of \mathcal{W}_2 input signals.
- We are currently trying to find a necessary and sufficient condition for finite-time stability (FTS)

(Again) Thank you!